

# Increase I<sup>2</sup>C or SMBus Data Rate and Reduce Power Consumption with Low Power Bus Accelerator

by Sam Tran

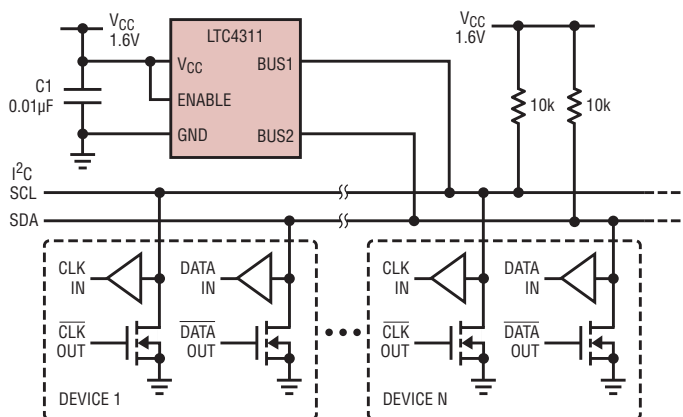
## Introduction

I<sup>2</sup>C and SMBus 2-wire buses use simple open-drain pull-down drivers with resistive or current source pull-ups. Communications protocols in these systems allow multiple devices to drive and monitor the bus without bus contention, creating a robust communications link. Unfortunately, as systems trend towards higher complexity and lower supply voltages, the advantages gained by the simplicity of the open-drain pull-down protocol are offset by the disadvantages of increased rise times and greater DC bus power consumption.

As designs require higher reliability and a greater number of features, the number of peripherals attached to the I<sup>2</sup>C or SMBus system increases. Some systems extend the bus to edge connectors where I/O cards with additional peripherals are removed and inserted onto the bus. The added peripherals directly increase the equivalent capacitance on the bus, slowing rise times. Slow rise times can seriously impact data reliability and limit the maximum practical bus speed to well below the established I<sup>2</sup>C or SMBus maximum transmission rate. Rise times can be improved by using lower bus pull-up resistor values or higher fixed current source values, but the additional bus pull-up current raises the low state bus voltage,  $V_{OL}$ , as well as the DC bus power consumption. Another issue in systems with swappable I/O cards is ESD susceptibility.

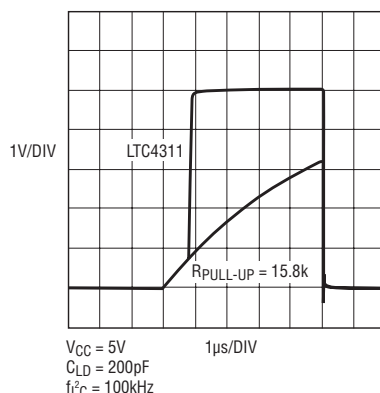
The LTC4311 bus accelerator addresses all of these issues. It comes in a tiny 2mm × 2mm DFN or SC70 package and operates over a wide power supply range of 1.6V to 5.5V, making it easy to fit in any number of applications.

Figure 1 shows a typical low voltage application circuit. The LTC4311



**Figure 1. Typical LTC4311 low voltage application circuit**

provides strong slew rate controlled pull-up currents on the bus for smooth, controlled transitions during rising edges to decrease rise times in highly capacitive systems, as shown in Figure 2. The LTC4311's slew rate controlled pull-up currents are strong enough to allow I<sup>2</sup>C or SMBus systems to achieve switching frequencies up to 400kHz for bus capacitances in excess of 1nF. In addition, because the accelerator pull-up impedance is significantly lower than the bus pull-up resistance, the system has greater immunity to noise on rising edges.



**Figure 2. Comparison of I<sup>2</sup>C waveform for the LTC4311 vs resistive pull-up**

The LTC4311's strong pull-up currents allow users to choose larger bus pull-up resistor values to reduce  $V_{OL}$ , DC bus power consumption and fall times, while still meeting rise time and switching frequency requirements. This is especially helpful for 2-wire systems where devices require resistances in series with their pull-down devices for ESD protection, since  $V_{OL}$  on these devices is reduced with larger bus pull-up resistor values. The larger bus pull-up resistor values are also beneficial in systems operating at bus supplies below 2.7V, where  $V_{OL}$  can be reduced well below the I<sup>2</sup>C specification, thereby increasing noise margins.

For I<sup>2</sup>C or SMBus systems where large numbers of I/O cards can be inserted and removed, the LTC4311's slew rate controlled pull-up currents properly address rise time issues despite large variations in bus capacitance. The controlled slew rate regulates the rise rate of the bus to 50V/µs–100V/µs, independent of bus capacitance.

With very light loads, as occurs when some or all cards are removed, no reflections occur on the bus due

to the slew rate controlled nature of the pull-up currents. When the bus is heavily loaded, the LTC4311 provides strong, controlled pull-up currents to significantly decrease rise times on the bus for capacitive loads well beyond 1nF.

All of these features, coupled with high  $\pm 8\text{kV}$  HBM ESD ruggedness, make the LTC4311 ideally suited, and in many cases necessary, for I<sup>2</sup>C or SMBus systems having large numbers of removable I/O cards.

### Circuit Operation

Figure 3 shows a functional block diagram of the LTC4311. The LTC4311 consists of two independent but identical circuits for each bus, consisting of a slew rate detector, two voltage comparators, and a slew rate controlled bus pull-up current.

The slew-rate detector monitors the bus and activates the accelerators only when the bus rise rate is greater than  $0.2\text{V}/\mu\text{s}$ . This ensures that the accelerators never turn on when the bus voltage is in a DC state or falling. The first voltage comparator is used to hold off the accelerator until the bus voltage exceeds a threshold voltage,  $V_{\text{THR}}$ . For supply voltages below  $2.7\text{V}$ ,  $V_{\text{THR}}$  is supply dependent, defined as  $0.3 \cdot V_{\text{CC}}$ . At higher supply voltages,  $V_{\text{THR}}$  is a constant  $0.8\text{V}$ . This optimizes the LTC4311 for use in low voltage systems, while offering rise time acceleration over a larger voltage range for I<sup>2</sup>C and SMBus systems operating at bus voltages above  $2.7\text{V}$ .

Once both conditions are met, the slew limited bus accelerator is enabled to quickly slew the bus. An internal slew rate comparator monitors the bus rise rate and controls the accelerator pull-up current to limit the bus rise rate to  $50\text{V}/\mu\text{s}$ – $100\text{V}/\mu\text{s}$ , independent of the bus capacitance. A second voltage comparator disables the pull-up current when the bus is within  $400\text{mV}$  of the bus pull-up supply.

For systems where a single bus accelerator is not sufficient to meet the rise time requirement, additional bus accelerators can be added in parallel to further decrease the rise time.

*continued on page 23*

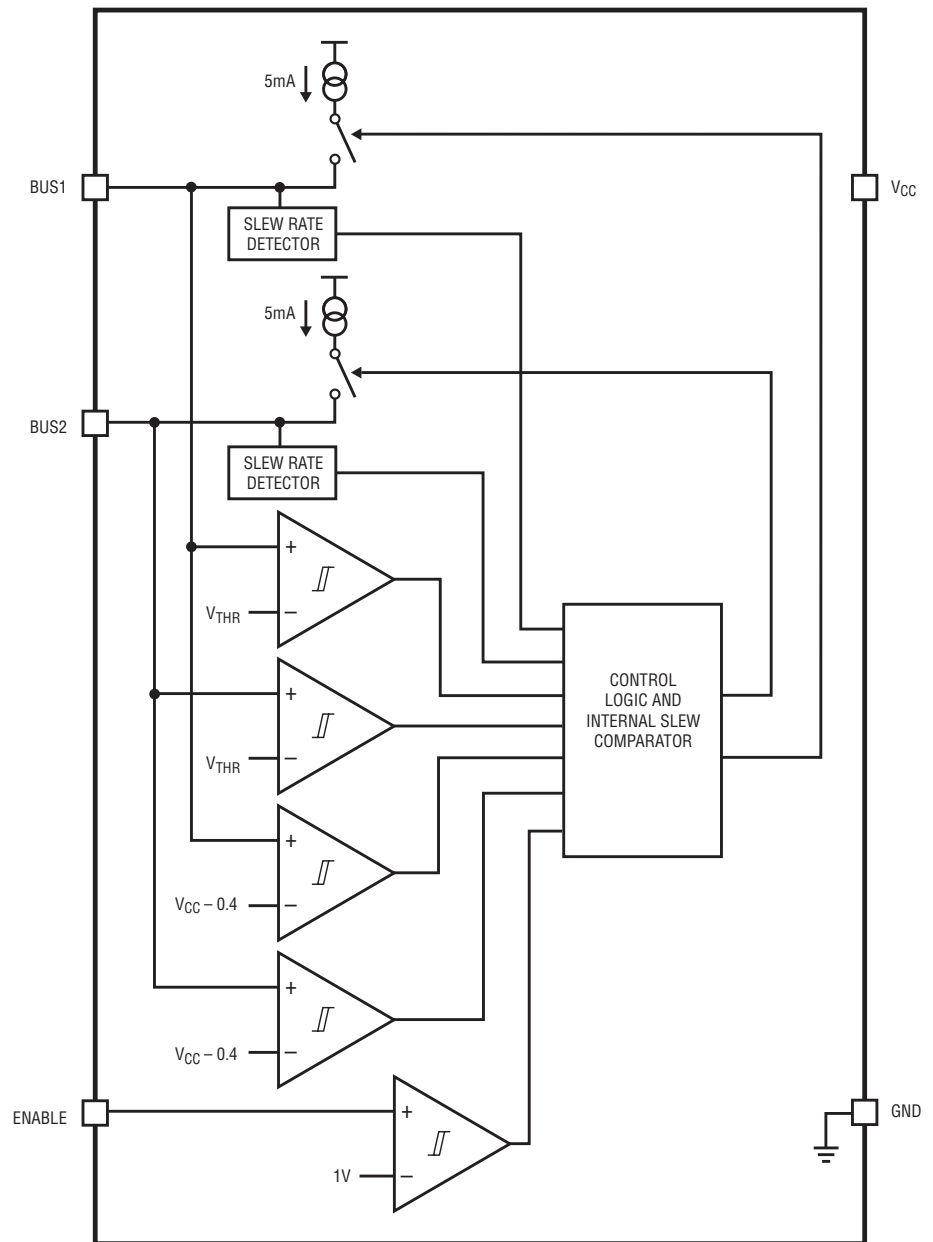


Figure 3. LTC4311 functional block diagram

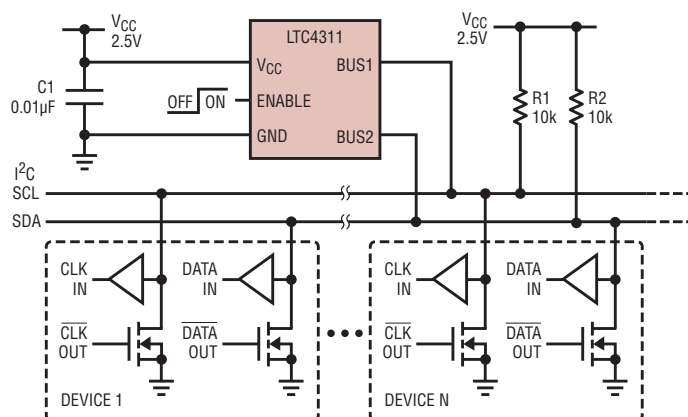


Figure 4. Typical LTC4311 application with low current shutdown

comparators incorporate anti-glitch circuitry. Any transient at the input of the monitor comparator must be of sufficient magnitude and duration (energy) to switch the comparator. Designs utilizing these single supervisors promote correct and glitch-free resets, which leads to stable and ultimately more reliable systems.

### Processor Communication

Two of the monitors (LTC2917 and LTC2918) communicate with host processors through their watchdog circuits. The basic requirement for the processor is to “pet” the watchdog periodically to avoid being “bitten” by the dog. Processor resets are invoked by the built-in watchdog hardware when the watchdog petting frequency has become too slow or too fast. Precise knowledge of the system’s timing characteristics is required to set the watchdog timeout period. Adjust the watchdog timeout period by connecting a capacitor between the watchdog timing input (WT) and ground. Connect WT to  $V_{CC}$  to achieve a default 1.6s timeout, without the need for external capacitance.

### Simple and Compliant Bias

A unique feature common to all four of these devices is the ability to provide operating bias from almost any positive voltage. It does not matter whether it is a 1.8V LDO, 5V switcher,

12V car battery, 24V wall-wart or 48V telecom supply; the integrated 6.2V shunt regulator can work with any system. For input voltages above 5.7V the only requirement is to size the bias resistor ( $R_{CC}$ ) to the range of the input voltage. Connect  $R_{CC}$  between the high voltage supply and the  $V_{CC}$  input. Below 5.7V, simply connect the supply directly to the  $V_{CC}$  input. Deriving resistor sizing for worst-case operation requires knowledge of the minimum ( $V_{S(MIN)}$ ) and maximum ( $V_{S(MAX)}$ ) input supply:

$$\frac{V_{S(MAX)} - 5.7V}{5mA} \leq R_{CC} \leq \frac{V_{S(MIN)} - 7V}{250\mu A}$$

Be sure to decouple the  $V_{CC}$  input using a 0.1 $\mu$ F (or greater) capacitor to ground.

### Qualify Once, Specify Forever

During product development cycles, power supply requirements often change. While supply requirements are changing, your choice of supervisor doesn’t have to. The LTC2915, LTC2916, LTC2917 and LTC2918 can relieve the burden of having to find the right supervisor for the job. Qualify any one of these parts and you can monitor any one of eight different supply voltages, each with three different internally fixed thresholds. You can also monitor any custom voltage down to 0.5V using an external resistor divider. Multi-supply monitoring is

easily achieved by using two or more devices and connecting their  $\overline{RST}$  outputs together.


### Meet Your Match

The LTC2915, LTC2916, LTC2917 and LTC2918 single supervisors are the perfect match for a variety of applications. Browse the applications shown in the figures and quickly find the right application for your system.

### Conclusion

The LTC2915, LTC2916, LTC2917 and LTC2918 are feature-laden single supervisors that can be comfortably placed near your monitored supply and/or microprocessor, leading to easy printed circuit board layout and reliable system operation.

Unprecedented configurability makes it possible to qualify and stock just one product that can meet all of your supervisory needs. Integration provides twenty-seven user-selectable monitor thresholds with  $\pm 1.5\%$  accuracy. Any non-standard threshold can be user-configured with the adjustable setting.

Other features include high voltage operation, configurable reset and watchdog timers, manual reset, and low quiescent current. External components are seldom required to realize fully functional designs. Electrical specifications are guaranteed from  $-40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . 

LTC4311, continued from page 9

### Auto Detect Standby Mode and Disable Mode

To conserve power, when both bus voltages are within 400mV of the bus pull-up supply, the LTC4311 enters standby mode, consuming only 26 $\mu$ A of supply current. When ENABLE is forced low, as shown in Figure 4, the LTC4311 enters a disable mode and consumes less than 5 $\mu$ A of supply current. Both bus pins are high impedance when in disable mode or when the LTC4311 is powered down, regardless of the bus voltage.

### Conclusion

The LTC4311 efficiently and effectively addresses slow rise times, decreased noise margins at low bus supplies, and increased DC bus power consumption found in 2-wire bus systems. Strong slew rate controlled pull-up currents quickly and smoothly slew the I<sup>2</sup>C or SMBus bus lines, decreasing rise times to allow up to 400kHz operation for bus capacitances in excess of 1nF. The advantages of the strong slew rate controlled currents extend to reducing the low state bus voltage,

DC bus power consumption, and fall times, since larger value bus pull-up resistors can be used.

With a small 2mm  $\times$  2mm  $\times$  0.75mm DFN or SC70 footprint, high  $\pm 8\text{kV}$  HBM ESD performance and low power consumption in standby or disable mode, the LTC4311 Low Voltage I<sup>2</sup>C or SMBus accelerator is also ideally suited for all I<sup>2</sup>C or SMBus systems. Examples of such systems include notebooks, palmtop computers, portable instruments, RAIDs, and servers where I/O cards are hot-swapped. 